S-DCCRN

SUPER WIDE BAND DCCRN WITH LEARNABLE COMPLEX FEATURE FOR SPEECH ENHANCEMENT

https://arxiv.org/pdf/2111.08387.pdf

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Research on different samplerate

- 8K Hz Narrow Band
- 16K Hz Wide Band
- 32K Hz Super Wide Band
- 44K Hz Full Band
- Most of the recent speech enhancement approaches mainly focus on wide-band signal with a sampling rate of 16K Hz.
- Research on super-wide-band or even full-band denoising is still lacked.

Challenges

- The challenges exist in modeling more frequency bands and particularly high frequency components.
 - How should we model more frequency bands?
 - How should we use the information in high frequency components?
- Modeling with larger dimensional features will cause higher complexity of the modeling.
 - How should we compress the dimension of feature ?

Answers of other approaches

HiFi-GAN

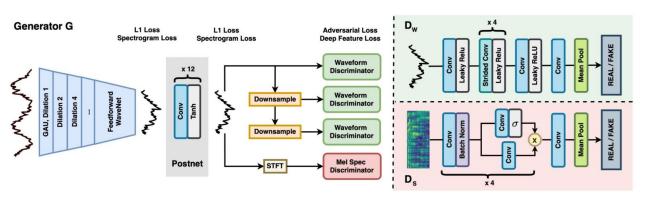
Use same network on different samplerate

RNNoise

• Bark scale: based on human perception

PercepNet

 Represent envelope from 0 to 20 kHz using 34 bands, spaced according to the human hearing equivalent rectangular bandwidth (ERB)



Bark scale

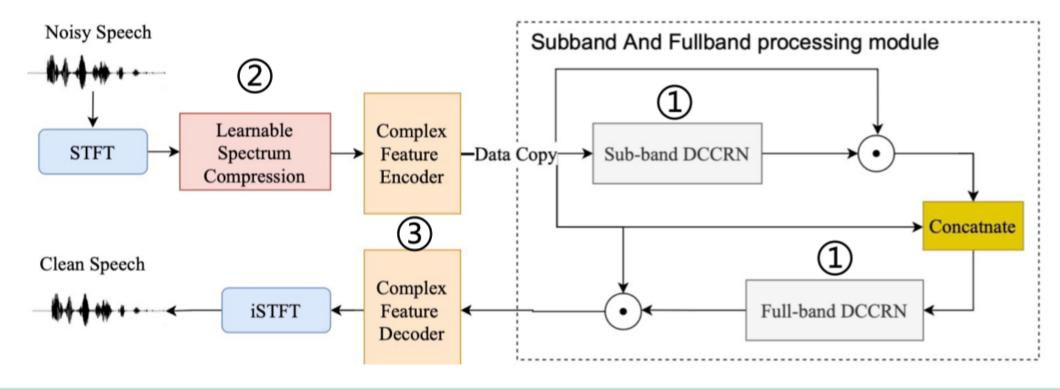
• The **Bark** scale is a psychoacoustical scale proposed by Eberhard Zwicker in 1961. It is named after Heinrich Barkhausen who proposed the first subjective measurements of loudness.

$$B = 13 an^{-1} \left(rac{0.76 f}{1000}
ight) + 3.5 an^{-1} \left(rac{f}{7500}
ight)^2$$

Critical Band (Bark)	Center Frequency (Hz)	Bandwidth (Hz)		
1	50	100		
2	150	100		
3	250	100		
4	350	100		
5	450	110		
6	570	120		
7	700	140		
8	840	150		
9	1000	160		
10	1170	190		
11	1370	210		
12	1600	240		
13	1850	280		
14	2150	320		
15	2500	380		
16	2900	450		
17	3400	550		
18	4000	700		
19	4800	900		
20	5800	1100		
21	7000	1300		
22	8500	1800		
23	10500	2500		
24	13500	3500		

https://www.prosoundtraining.com/2019/07/26/why-equalize-in-1-3-octave-bands/

Answer of S-DCCRN



#	S-DCCRN 贡献	动机
1	Sub-band & Full-band DCCRN (SAF)	首先子带DCCRN精细化学习高低频信息,然后全带DCCRN结合 高低频信息,起到平滑衔接作用
2	Learnable Spectrum Compression (LSC)	通过网络学习, 动态调整不同频带能量
3	Complex Feature Encoder/Decoder (CFE/CFD)	在同16K降噪模型保持相同的较低频率分辨率的同时,通过复数特征编码从谱上获取跟多信息

Complex Feature Encoder/Decoder

- 1 extract high dimensional information
- 2 capture long-term contextual features from time scale
- ③ extract complex local features
- 4 process the estimated real/imag features
- 5 pixel convolution is considered as a better alternative for transposed convolution to avoid checkerboard artifacts
- 6 revert the high-dimensional feature to the time-frequency domain

The STFT length is 512.
The hidden channels of the complex feature encoder/decoder module are 32.

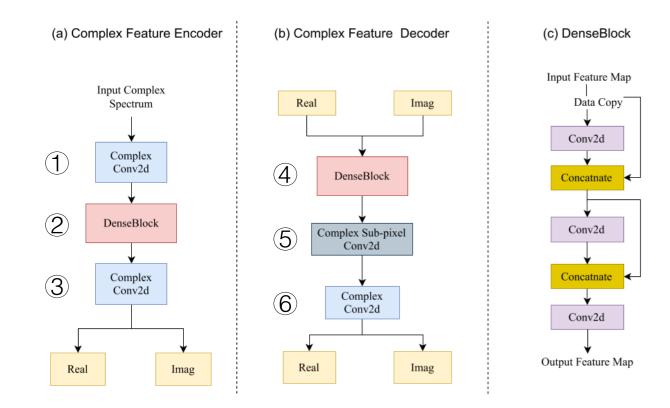


Fig. 2. Complex feature encoder/decoder (CFE/CFD) module

Sub-band and Full-band Processing Module

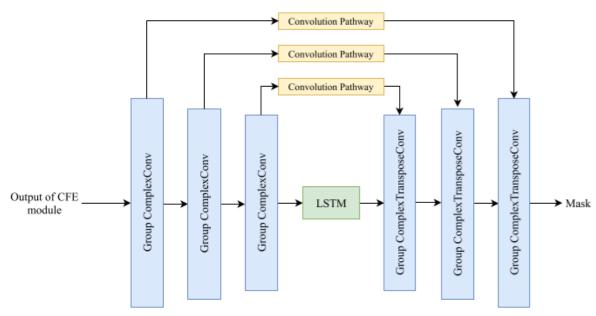


Fig. 3. The sub-band DCCRN module

The skip pathway between encoder and decoder consists of a complex convolution block and batch normalization.

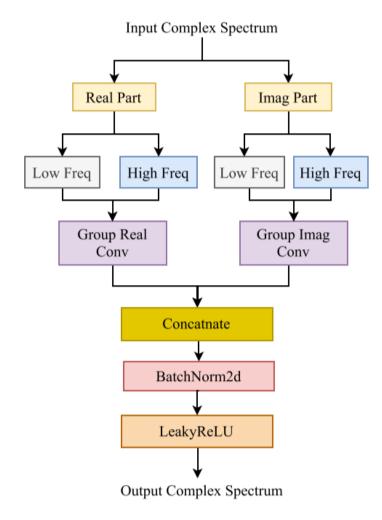


Fig. 4. Complex group convolution

Learnable Spectrum Compression

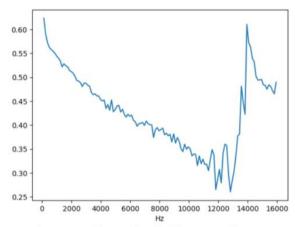


Fig. 5. Compression ratio of different frequency automatically learned by the proposed learnable spectrum compression.

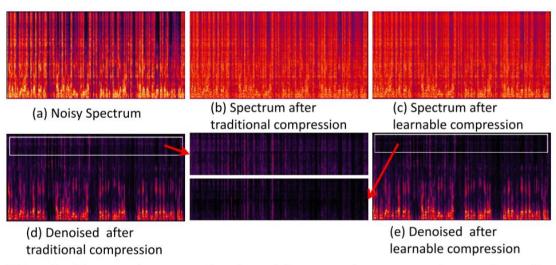


Fig. 6. Comparison on the denoising result on a testing noisy clip for the cases with/without learnable spectrum compression.

In detail, the learnable spectrum compression

can be described as

$$Y_{LSC} = |Y|^{\alpha} e^{j\varphi_Y}$$

 $Y_{\rm LSC} = |Y|^{\alpha} e^{j\varphi_Y}$ where Y and α denote the noisy spectrum and the learnable parameters respectively.

Loss Function

- SI-SNR loss in time-domain
 - scale-invariant source-to-noise ratio

$$\begin{cases} \mathbf{s}_{target} := \frac{\langle \hat{\mathbf{s}}, \mathbf{s} \rangle \mathbf{s}}{\|\mathbf{s}\|^2} \\ \mathbf{e}_{noise} := \hat{\mathbf{s}} - \mathbf{s}_{target} \\ \text{SI-SNR} := 10 \log_{10} \frac{\|\mathbf{s}_{target}\|^2}{\|\mathbf{e}_{noise}\|^2} \end{cases}$$

Complex mean-squared error (MSE) loss in T-F domain

$$\mathcal{L}_{\text{cMSE}} = \frac{1}{T \times F} \sum_{t,f} \left| |X| e^{j\varphi_X} - |\hat{X}| e^{j\varphi_{\hat{X}}} \right|$$

Kullback-Leibler Divergence in T-F domain

$$\mathcal{L}_{\text{KL}} = \frac{1}{T \times F} \sum_{t,f} \hat{X} \cdot log(\frac{\hat{X}}{X})$$

Result

Table 1. Results of various models and ablation experiments on Voice Bank and DEMAND set.

Model	# Para.(M)	PESQ	CSIG	COVL	CBAK	STOI
Noisy	-	1.97	3.35	2.63	2.44	0.921
RNNoise	0.06	2.34	3.40	2.84	2.51	0.922
PercepNet	8	2.73	-	-	-	-
DCCRN	3.7	2.54	3.74	3.13	2.75	0.938
SP	2.76	2.63	3.86	3.23	3.03	0.935
SAF	2.73	2.71	3.94	3.31	3.08	0.937
+ SC	2.73	2.76	3.98	3.36	2.87	0.938
+ LSC	2.73	2.77	3.98	3.35	2.92	0.938
+ CFE/CFD	2.34	2.69	3.90	3.28	3.08	0.939
+ SC	2.34	2.77	3.98	3.37	2.87	0.940
+ LSC (S-DCCRN)	2.34	2.84	4.03	3.43	2.97	0.940

Table 2. MOS and DNSMOS results on DNS-2021 blind test set.

Model	MOS	DNSMOS*
Noisy	1.66	2.94
RNNoise	2.32	3.07
DCCRN	3.30	3.31
SAF	3.20	3.33
S-DCCRN	3.62	3.43

^{*:} Calculated on downsampled speech (16K Hz)

Demo: https://imybo.github.io/S-DCCRN/